

SC2603A Simple PWM Boost Converter In Small SOT23-6 Package

POWER MANAGEMENT

Description

The SC2603A is a versatile, low-cost, voltage-mode PWM controller designed for DC/DC power supply applications. A simple, fixed-voltage boost regulator can be implemented using the SC2603A with a minimum of external components. The small device footprint allows for compact circuit design.

SC2603A features include a temperature compensated voltage reference, triangle wave oscillator, current limit comparator, frequency shift over-current protection, and an internally compensated error amplifier. Pulse by pulse current limiting is implemented by sensing the differential voltage across an external resistor, or an appropriately sized PC board trace.

The SC2603A operates at a fixed frequency of 200kHz, providing an optimum compromise between efficiency, external component size, and cost.

Features

- Low cost / small size
- Input voltage range: 5V to 30V
- 200kHz switching frequency
- Up to 95% efficiency
- 1% reference voltage accuracy
- Typical 800mA PWM gate drive capability
- Internal compensation
- Over current protection
- SOT23-6 package is fully WEEE and RoHS compliant

Applications

- Portable Devices
- Flat Panel TV
- TV Set Top Box
- Auxiliary Supplies
- Peripheral Card Supplies
- Industrial Power Supply
- High Density DC/DC Conversion



Typical Application Circuit

SC2603A

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Absolute Maximum Rating

Exceeding the specifications below may result in permanent damage to the device, or device malfunction. Operation outside of the parameters specified in the Electrical Characteristics section is not implied. Exposure to Absolute Maximum rated conditions for extended periods of time may affect device reliability.

Parameter	Symbol	Maximum	Units
Input Voltage	V _{IN} to GND	-0.3 to 20	V
GATE Pin Voltage	V _{GATE}	-0.3 to 20	V
EN/SS Pin Voltage	V _{EN/SS}	-0.3 to 7	V
FB Pin Voltage	V _{FB}	-0.3 to 7	V
Thermal Resistance Junction to Ambient ⁽¹⁾	θ _{JA}	165	°C/W
Thermal Resistance Junction to Case	θ _{JC}	102	°C/W
Maximum Junction Temperature	TJ	150	°C
Storage Temperature Range	T _{STG}	-45 to +150	°C
Lead Temperature (Soldering)10 sec	T _{LEAD}	300	°C
ESD Rating (Human Body Model)	ESD	2	kV

Note:

(1) Mounting pad has to be larger than one square inch on two sided printed circuit board with minimum 1 ounce copper

Electrical Characteristics

Unless specified: V_{IN} = 12V, V_{O} = 25V; T_{A} = T_{J} = -40 °C to 125 °C

Parameter	Symbols	Conditions	Min	Тур	Max	Units	
General	General						
V _{IN} Supply Voltage	V _{IN}		4.5		16	V	
V _{IN} UVLO Threshold	V _{UVLO}	V _{IN} ramping up		4.2	4.5	V	
V _{IN} UVLO Hysteresis	V _{HYS}			400		mV	
V _{IN} Supply Current		Current into V_{IN} pin during switching		6.0	9.0	mA	
V _{IN} Shutdown Current		EN/SS=Low		5.0	6.0	mA	
Error Amplifier						·	
Reference Voltage	V _{FB}		1.225	1.250	1.275	V	
Feedback Bias Current	I _{FB}	V_{IN} = 12V, V_{FB} = $V_{EN/SS}$		2.0	8.0	μΑ	
Error Amplifier Gain(1)				105		V/V	
Oscillator						·	
Oscillator Frequency		T _A = 25 °C	190	200	210	kHz	
Oscillator Frequency Shift		V _{FB} < V _{REF} /2		50		kHz	
Maximum Duty Cycle			90	95		%	
Internal Ramp Peak	V _{R_PEAK}			1.4		V	
Internal Ramp Valley	V _{R_VALLEY}			0.4		V	

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Electrical Characteristics (Cont.)

Parameter	Symbols	Conditions	Min	Тур	Max	Units	
Regulation	Regulation						
Load Regulation		$I_{o} = 0.1A \text{ to } 1A$		0.3	0.5	%	
Line Regulation		$V_{IN} = 5V$ to 15V, $I_{o} = 0.1A$		0.5	1.0	%	
PWM Gate Drive							
GATE Source Current		V _{IN} = 12V, C _{GATE} = 10nF	0.5	0.8	1.1	A	
GATE Sink Current		V _{IN} = 12V, C _{GATE} = 10nF	0.5	0.8	1.1	A	
Enable and Soft Start			·			·	
Soft Start Charge Current	I _{ss}			60		μA	
Enable/Soft Start Threshold		Pull down below this level to disable PWM Switch gate	0.09	0.12		V	
Switch Current Limit							
Current Limit Threshold		CS	65	73	80	mV	

Note:

(1) Guaranteed by design

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(6 Pin SOT-23)

Ordering Information

Part Number	Package
SC2603ASKTRT ^(1,2)	SOT23-6
SC2603AEVB	Evaluation Board

Notes:

(1) Only available in tape and reel packaging. A reel contains 3000 devices.

(2) Lead free product. This product is fully WEEE and RoHS compliant.

Pin Descriptions

Pin	Pin Name	Pin Function	
1	FB	Error amplifier inverted input	
2	GND	Device ground	
3	EN/SS	Enable / Soft Start Pin	
4	GATE	PWM gate driver output	
5	VIN	Device input voltage	
6	CS	Current Sense input	

Block Diagram





Applications Information

Theory of Operation

The SC2603A is a versatile, low-cost, voltage-mode PWM controller designed for DC/DC power supply applications. In normal condition, the SC2603A operates at a fixed 200kHz. One exception is that when the FB pin voltage drops to one half, the frequency will be shifted to one fourth of 200kHz.

SC2603A features include a temperature compensated voltage reference, triangle wave oscillator, current limit comparator, frequency shift over-current protection, and an internally compensated error amplifier. Pulse by pulse current limiting is implemented by sensing the differential voltage across an external resistor, or an appropriately sized PC board trace.

Setting the Output Voltage

As shown in Fig.1, an external resistive divider R2 and R4 with its center tap tied to the FB pin sets the output voltage.

$$\mathsf{R}_2 = \mathsf{R}_4 \left(\frac{\mathsf{V}_{\mathsf{OUT}}}{1.25\mathsf{V}} - \mathsf{1} \right)$$

Soft-Start

Soft-start prevents a DC-DC converter from drawing excessive current from the power source during start up. If the soft-start time is made sufficiently long, then the output will enter regulation without overshoot. An external capacitor is connected from the EN/SS pin to the ground. After the part's input voltage surpasses the UVLO, the EN/SS pin is allowed to charge its capacitor with 60µA. When 100mV is reached at the EN/SS pin, the internal comp node is at 0.4V. The SC2603A will start switching, and the converter output comes into regulation. Because the FB pin voltage follows EN/SS pin voltage, finally, the EN/SS pin voltage will be clamped at 1.25V same as the reference voltage.

Under Voltage Lockout

The under voltage lockout circuit of the SC2603A assures that the Gate output remains in the off state whenever the supply voltage drops below set parameters. Lockout occurs if V_{IN} falls below 3.8V. Normal operation resumes once Vin rises above 4.2V.

Maximum Duty Cycle

The maximum duty cycle, $\mathsf{D}_{_{\rm MAX}}$ defines the upper limit of power conversion ratio

$$\frac{V_{\text{OUT}}}{V_{\text{IN}}} = \frac{1}{1 - D_{\text{MAX}}}$$

Clamping Diode D2 at VIN Pin

Since the V_{IN} pin of the SC2603A is rated at 20V, when the supply voltage is higher than 16V, a Zener diode D2 is required to limit the SC2603A supply voltage, as shown in Fig.1. The clamping voltage of D2 has to be not higher than 15V.

Over Current Protection (OCP)

The over-current protection is implemented by a sensing resistor and an internal current-limit comparator with a threshold of 70mV. In each switching cycle, if the switching current through the sensing resistor causing the voltage at the CS pin reaches 70mV, the SC2603A will immediately turn the gate pulse off to limit the current through the power switch.

The figure below shows the current waveforms run into over-current protection. At t1, the input current lin begins to increase as the load lout builds up until it hits the OCP threshold at t2. At t2, Vout begins to drop as lout starts increasing toward t3 while Vin and lin remain the same. At t3, there two events are happening. 1) Vout is at half of its regulating voltage. At this point, the internal oscillator runs into frequency shift mode and shifts the frequency to one fourth of the operating frequency. 2) The diode starts conducting current without limit causing Vin to fall due to supply current limit. Finally, Vin and Vout both drops to zero and lin becomes the input supply current limit.

A RC filter at the CS pin is necessary to filter out noise to ensure accurate sensing.

The value of the sensing resistor can be calculated by the following equation:

Rs = 70mV/Ipeak



Applications Information (Cont.)

where Ipeak is the input peak current through the power MOSFET.





Layout Guidelines

Careful attentions to layout requirements are necessary for successful implementation of the SC2603A PWM controller. High currents switching at 200kHz are present in the application and their effect on ground plane voltage differentials must be understood and minimized.

1) The high power parts of the circuit should be laid out first. A ground plane should be used. The number and position of ground plane interruptions should be such as to not unnecessarily compromise ground plane integrity. Isolated or semi-isolated areas of the ground plane may be deliberately introduced to constrain ground currents to particular areas, for example the input capacitor and bottom Schottky ground.

2) The loop formed by the output Capacitor(s) (Cout), the FET (Q1), the current sensing resistor, and the Schottky (D1) must be kept as small as possible, as shown on the layout diagram in Fig.4. This loop contains all the high current, fast transition switching. Connections should be as wide and as short as possible to minimize loop inductance. Minimizing this loop area will reduce EMI, lower ground injection currents, resulting in electrically "cleaner" grounds for the rest of the system and minimize source ringing, resulting

in more reliable gate switching signals.

3) The connection between the junction of Q1, D1 and the output capacitor should be a wide trace or copper region. It should be as short as practical. Since this connection has fast voltage transitions, keeping this connection short will minimize EMI.

4) The Output Capacitor(s) (Cout) should be located as close to the load as possible, fast transient load currents are supplied by Cout only, and connections between Cout and the load must be short, wide copper areas to minimize inductance and resistance.

5) The SC2603A is best placed over an isolated ground plane area. The soft-start capacitor, the Vin decoupling capacitor, and the current sensing filter capacitor should also connected to this ground pad area. This isolated ground area should be connected to the main ground by a trace that runs from the GND pin to the ground side of the output capacitor. If this is not possible, the GND pin may be connected to the ground path between the Output Capacitor and the Cin, Q1, D1 loop. Under no circumstances should GND be returned to a ground inside the Cin, Q1, D1 loop.

6) Input voltage of the SC2603A should be supplied from the power rail through a 2.2~10 Ω resistor, the Vin pin should be decoupled directly to GND by a 0.1μ F~1 μ F ceramic capacitor, trace lengths should be as short as possible.

7) An RC low pass filter is required to filter out leading edge current spikes on the current sensing resistor. The filter capacitor needs to be very close to the CS and GND to reduce noise pickup.





Applications Information (Cont.)

Layout diagram for the SC2603A







Applications Information (Cont.)

Typical application schematic with 12V input and 24V/1A output



Bill of materials

Item	Quantity	Reference	Part	Vendor
1	1	C1	220uF/16V	Panasonic
2	1	C2	330uF/35V	Nichicon, VR
3	2	C3,C10	1uF/16V	Vishay
4	1	C4	0.1uF	Vishay
5	1	C5	120pF	Vishay
6	1	C11	1uF/50V	Vishay
7	1	Cc	56nF	Vishay
8	1	D1	CMSH2-40M	Central Semi
9	1	L1	33uH/5.1A (HC8-330)	Coiltronics
10	1	Q1	IRF7821	IR
11	1	R1	2R2	Vishay
12	1	R2	226k	Vishay
13	1	R4	12.4k	Vishay
14	1	R5	1k	Vishay
15	1	R6	20m	Vishay
16	1	Rc	2.3k	Vishay
17	1	U1	SC2603A	Semtech



Outline Drawing - SOT23-6



Land Pattern - SOT23-6



	DIMENSIONS				
DIM	INCHES	MILLIMETERS			
С	(.098)	(2.50)			
G	.055	1.40			
Р	.037	0.95			
X	.024	0.60			
Y	.043	1.10			
7	141	3.60			

NOTES

1. THIS LAND PATTERN IS FOR REFERENCE PURPOSES ONLY. CONSULT YOUR MANUFACTURING GROUP TO ENSURE YOUR COMPANY'S MANUFACTURING GUIDELINES ARE MET.

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